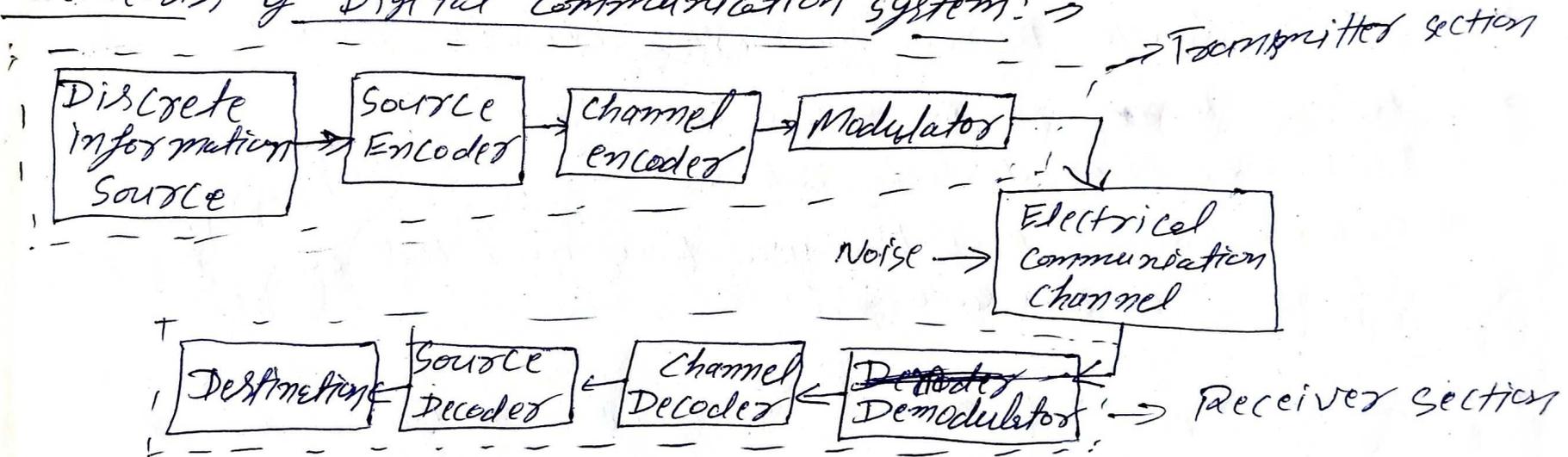


Digitized Communication System: Digital communication is that type of communication in which the message or information signal to be transmitted is digital in nature.

Elements of Digital communication system:



Information source:

Generates the message signal to be transmitted  
 there are two types of information source.

- (a) Analog information source.
- (b) ~~Digi~~ Discrete information source.

## Source encoder\* and Decoder\*\* ⇒

- \* Converts the symbols produced by the information source into digital form (i.e. Sequence of 1's and 0's)
- \* Every binary '1' and '0' is called a bit and the group of bits is called a code word.

\*\* Source decoder perform the reverse operation of source encoder.

\* Important Parameters for source encoder are

→ Block size (max. distinct code word)

→ code word length (No. of bits used to represent each code) ( $2^n$  bits).

→ data rate (0/p bits per second)

channel encoder and Decoder ⇒ It is used to prevent the channel from noise. channel may be wireless or wired. the prevention of channel from noise is done by adding an extra bit in the source encoder for making even parity. For example ⇒

Output of source encoder	Bits to be added by the channel encoder for even parity (extra bit) $b_0$	output of a channel encoder
$b_3$ $b_2$ $b_1$		$b_3$ $b_2$ $b_1$ $b_0$
0 0 0	0	0 0 0 0
0 1 0	1	0 1 0 1
0 1 1	0	0 1 1 0

## Source encoder\* and Decoder\*\* ⇒

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- \* Every binary '1' and '0' is called a bit and the group of bits is called a code word.

\*\* Source decoder perform the reverse operation of source encoder.

\* Important Parameters for source encoder are

→ Block size (max. distinct code word)

→ code word length (No. of bits used to represent each code) ( $2^n$  bits).

→ data rate (0/1 bits per second)

channel encoder and Decoder ⇒ It is used to prevent the channel from noise. channel may be wireless or wired. the prevention of channel from noise is done by adding an extra bit in the source encoder for making even parity. For example ⇒

Output of source encoder	Bits to be added by the channel encoder for even parity	output of a channel encoder
$b_3$ $b_2$ $b_1$	(extra bit) $b_0$	$b_3$ $b_2$ $b_1$ $b_0$
0 0 0	0	0 0 0 0
0 1 0	1	0 1 0 1
0 1 1	0	0 1 1 0

## Digital Modulator & Demodulator :->

- \* A high frequency continuous sinusoidal wave is used as a carrier signal in digital modulation.
- \* the amplitude, phase, or frequency of carrier signal is varied in accordance with the input binary sequence.

Channel :-> Channel is used to transmit the data from one point to another point. Channel may be wired or wireless. The main parameters of channel are as follows.

- (a) Signal Attenuation.
- (b) Noise
- (c) Multipath distortion
- (d) Amplitude and phase distortion.

## Advantages of Digital Communication System :->

- (a) Simpler and cheaper because of the advances made in IC technology.
- (b) Speech, video and other data may be merged and transmitted over a common channel using multiplexing.
- (c) Using data encryption, only permitted receivers may be allowed to detect the transmitted data.

(d) Due to channel coding, the error may be detected and corrected at the receiver's.

### Disadvantages of Digital Comm. system:

- (a) Because of analog to digital conversion, the data rates becomes high, hence more transmission bandwidth is required for digital communication.
- (b) Needs synchronization.

Line coding:  $\rightarrow$  The digital data can be transferred by various lines codes such as ON-OFF, Polar, bipolar etc. This is called line coding.

### Properties of line coding:

- (a) Transmission Bandwidth  $\rightarrow$  Must be as small as possible.
- (b) Power efficiency:  $\rightarrow$  Transmitted power for a line code should be as small as possible.

- (c) - Error Detection and correction capability:  $\Rightarrow$  It must be possible to detect and correct errors.
- (d)  $\Rightarrow$  Adequate Timing content:  $\Rightarrow$  It must be possible to extract timing or clock information from the signal.
- (e)  $\Rightarrow$  Transparency:  $\Rightarrow$  The wave form of the line code should be transparent to the digital data being transmitted.
- (f)  $\Rightarrow$  The line code should protect the channel from noise and other ~~information~~ interference.

Formats of line code  $\Rightarrow$  (a) Unipolar RZ Format  $\Rightarrow$  (Return to zero)

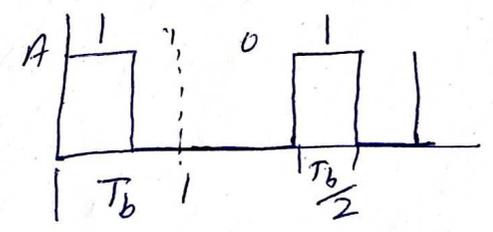
If symbol '1' is transmitted then

$$x(t) = \begin{cases} A & \text{for } 0 \leq t \leq \frac{T_b}{2} \\ 0 & \text{for } \frac{T_b}{2} \leq t \leq T_b \end{cases} \quad \text{where } T_b \text{ is the time period of waveform}$$

For symbol '0'

$$x(t) = \begin{cases} 0 & \text{for } 0 \leq t \leq T_b \end{cases}$$

Unipolar RZ



b) Unipolar NRZ Format  $\Rightarrow$

for symbol '1'  $x(t) = \begin{cases} A & \text{for } 0 \leq t \leq T_b \end{cases}$

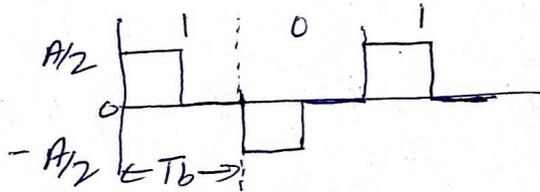
for symbol '0'  $x(t) = \begin{cases} 0 & \text{for } 0 \leq t \leq T_b \end{cases}$

c) Polar RZ Format  $\Rightarrow$

for symbol '1'  $x(t) = \begin{cases} A/2 & \text{for } 0 \leq t \leq T_b/2 \\ 0 & \text{for } T_b/2 \leq t \leq T \end{cases}$

for symbol '0'  $x(t) = \begin{cases} -A/2 & \text{for } 0 \leq t \leq T_b/2 \\ 0 & \text{for } T_b/2 \leq t \leq T \end{cases}$

Polar RZ

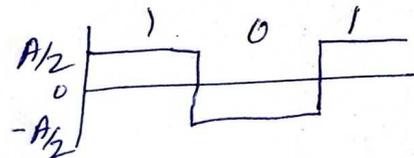


d) Polar NRZ Format :-

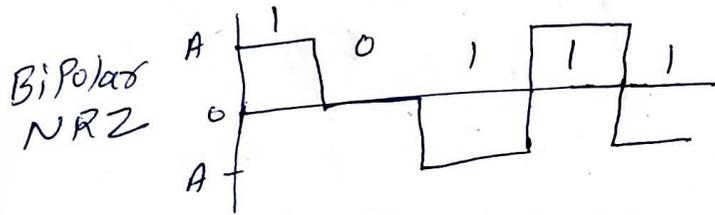
for symbol '1'  $x(t) = \begin{cases} A/2 & \text{for } 0 \leq t \leq T_b \end{cases}$

for symbol '0'  $x(t) = \begin{cases} -A/2 & \text{for } 0 \leq t \leq T_b \end{cases}$

Polar NRZ



- (E) BiPolar NRZ Format (Alternate Mark Inversion i.e. AMI)
- \* Successive '1' are represented by pulses with alternate Polarity.
  - \* '0' are represented by No Pulse.

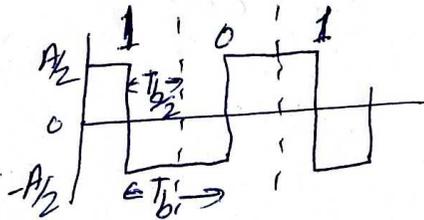


- (F) Split Phase Manchester →

for symbol '1'  $x(t) = \begin{cases} A/2 & \text{for } 0 \leq t < T_b/2 \\ -A/2 & \text{for } T_b/2 \leq t < T_b \end{cases}$

for symbol '0'  $x(t) = \begin{cases} -A/2 & \text{for } 0 \leq t < T_b/2 \\ A/2 & \text{for } T_b/2 \leq t < T_b \end{cases}$

Split Phase  
Manchester



(G1) Quaternary NRZ format is used to reduce the bandwidth. Here the combination of <sup>two</sup> bits are transmitted. It is also known as Polar Quaternary Format.

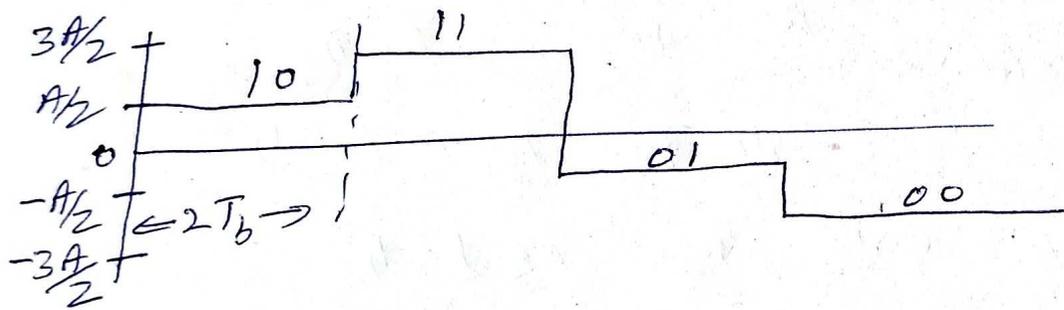
Message combination

Signal Amplitude

00  
01  
10  
11

$-3\frac{A}{2}$   
 $-A/2$   
 $A/2$   
 $3\frac{A}{2}$

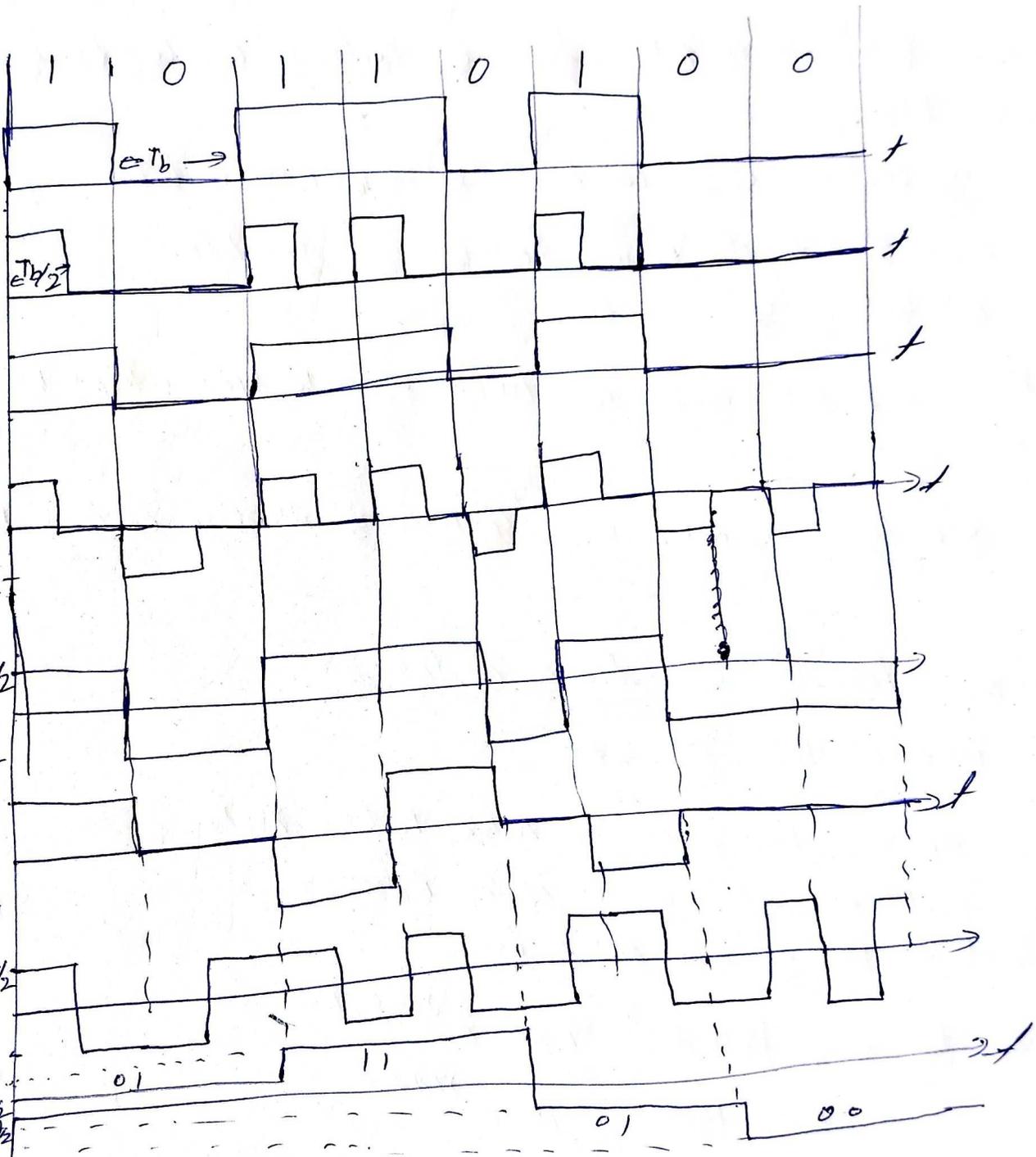
For duration of  $2T_b$



Q → Draw the line coding of given sequence.

10110100

Solu<sup>ns</sup>



Binary

UniPolar RZ

UniPolar NRZ

Polar RZ

Polar NRZ

Bipolar NRZ

Split Phase Manchester

$3A/2$

$-3A/2$

## M-ary Coding

- \* In polar quaternary NRZ type of coding, we combine two successive bits.
- \* In M-ary coding we combine 'K' successive bits.
- \* For  $K=2$ ,  $2^2=4$  distinct levels are possible. This is called 4-ary coding.
- \* For  $K=3$ ,  $2^3=8$  distinct levels are possible. This is called 8-ary coding.
- \* Polar quaternary NRZ is also type of M-ary coding for  $K=1$ .

## Power Spectral Density of Line Codes

(a) PSD for NRZ Unipolar format

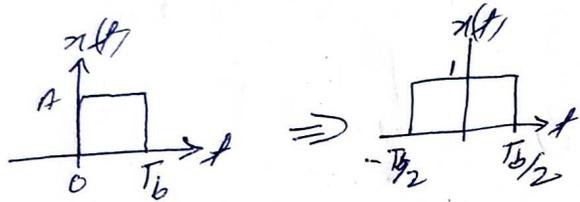
Step-1  $\Rightarrow$  Find Fourier transform of NRZ pulse  $x(t)$ .

Step-2  $\Rightarrow$  Find autocorrelation of unipolar  $R_c(m)$ .

Step-3  $\Rightarrow$  Calculate PSD of  $x(t)$  given by.

Winer  $\Rightarrow S(f) = \frac{1}{T} |X(f)|^2 \sum_{n=-\infty}^{\infty} R_c(m) e^{-j2\pi f_n T}$   
Kintchine  
where  $H(f) = \int_{-\infty}^{\infty} x(t) e^{-j2\pi f t} dt$

For example: Calculate NRZ Pulse Fourier transform:-



Step-1 → Find Fourier transform by

$$X(f) = \int_{-\infty}^{\infty} x(t) e^{-j2\pi f t} dt$$

$$= \int_{-T_b/2}^{T_b/2} (1) e^{-j2\pi f t} dt = \left[ \frac{e^{-j2\pi f t}}{-j2\pi f} \right]_{-T_b/2}^{T_b/2}$$

$$= -\frac{1}{j2\pi f} \left[ e^{-j2\pi f T_b/2} - e^{j2\pi f T_b/2} \right]$$

$$= \frac{1}{j2\pi f} \left[ e^{j2\pi f T_b/2} - e^{-j2\pi f T_b/2} \right]$$

$$= \frac{1}{\pi f} \left[ \frac{e^{j\pi f T_b} - e^{-j\pi f T_b}}{2j} \right]$$

we know that

$$\sin \theta = \frac{e^{j\theta} - e^{-j\theta}}{2j}$$

$$X(f) = \frac{1}{\pi f} \sin(\pi f T_b)$$

we know that the sinc form

$$\text{Sinc}(x) = \frac{\sin(\pi x)}{\pi x}$$

$$X(f) = T_b \frac{\sin(\pi f T_b)}{\pi f T_b}$$

$$X(f) = T_b \text{Sinc}(f T_b)$$

New Step-II  $\rightarrow$  Calculation of Auto correlation term.

for unipolar NRZ format

$$A_k = \left. \begin{array}{l} A \quad \text{Symbol } 1 \\ 0 \quad \text{Symbol } 0 \end{array} \right\}$$

Probability of occurrence.  $P(A_k=0) = P(A_k=A) = \frac{1}{2}$

Auto correlation of  $A_k$  with  $A_k$  is given as.

$$R_p(m) = E[A_k \cdot A_{k-m}]$$

$$\text{for } m=0; R_p(0) = E[A_k A_k] = E[A_k^2]$$

~~we~~ we know the mean value of discrete random variable is

$$m_x = E(x) = \bar{x} = \sum_{i=1}^n x_i P(x_i)$$

$$\text{and } E(x^2) = \sum_{i=1}^n x_i^2 P(x_i) \quad \text{--- (1)}$$

formation of table:

$A_k$	$A_k$	Amplitude $A_k$ $A_k$	$A_k^2$	$P(x_i)$
0	0	0 0	0	$\frac{1}{2}$
1	1	A A	$A^2$	$\frac{1}{2}$

$$\text{from eqn (1)} \quad E(x^2) = 0 \times \frac{1}{2} + A^2 \times \frac{1}{2} = \frac{A^2}{2}$$
$$E(A_k^2) = \frac{A^2}{2}$$

Now for  $n \neq 0$   $R_q(m) = E[A_k \cdot A_{k-n}]$

formation of table.

$A_k$	$A_{k-n}$	Amplitude		$A_k \cdot A_{k-n}$	$P(A_k \cdot A_{k-n})$
		$A_k$	$A_{k-n}$		
0	0	0	0	0	$\frac{1}{4}$
0	1	0	A	0	$\frac{1}{4}$
1	0	A	0	0	$\frac{1}{4}$
1	1	A	A	$A^2$	$\frac{1}{4}$

$$\text{Now } E[A_k \cdot A_{k-n}] = \sum_{i=0}^n x_i^2 p_{x_i}$$

$$= 0^2 \cdot \frac{1}{4} + 0^2 \cdot \frac{1}{4} + 0^2 \cdot \frac{1}{4} + A^2 \cdot \frac{1}{4}$$

$$E[A_k \cdot A_{k-n}] = \frac{A^2}{4}$$

$$R_q(n) = \begin{cases} \frac{A^2}{2} & \text{for } n=0 \\ \frac{A^2}{4} & \text{for } n \neq 0 \end{cases}$$

Now step-III  $\Rightarrow$  PSD of unipolar NRZ

$$S(f) = \frac{1}{T_b} |X(f)|^2 \sum_{n=-\infty}^{\infty} R_q(n) e^{-j2\pi f n T_b}$$

Now put the value of  $X(f)$  and  $R_2(m)$

~~$S(f)$~~

$$S(f) = \frac{1}{T_b} [T_b^2 \text{sinc}^2(fT_b)] \left[ \sum_{n=0}^{\infty} \frac{A^2}{2} e^{-j2\pi f n T_b} + \sum_{\substack{n=-\infty \\ n \neq 0}}^{\infty} \frac{A^2}{4} e^{-j2\pi f n T_b} \right]$$

$$S(f) = \frac{A^2 T_b}{2} \text{sinc}^2(fT_b) + \frac{T_b A^2}{4} \text{sinc}^2(fT_b) \left[ \sum_{\substack{n=-\infty \\ n \neq 0}}^{\infty} e^{-j2\pi f n T_b} \right]$$

$$S(f) = \frac{A^2 T_b}{2} \text{sinc}^2(fT_b) + \frac{T_b A^2}{4} \text{sinc}^2(fT_b) \left[ \sum_{n=-\infty}^{\infty} e^{-j2\pi f n T_b} - 1 \right]$$

$$= \frac{A^2 T_b}{2} \text{sinc}^2(fT_b) - \frac{T_b A^2}{4} \text{sinc}^2(fT_b) + \frac{T_b A^2}{4} \text{sinc}^2(fT_b) \sum_{n=-\infty}^{\infty} e^{-j2\pi f n T_b}$$

$$= A^2 T_b \text{sinc}^2(fT_b) \left[ \frac{1}{2} - \frac{1}{4} \right] + \frac{T_b A^2}{4} \text{sinc}^2(fT_b) \sum_{n=-\infty}^{\infty} e^{-j2\pi f n T_b}$$

$$S(f) = \frac{A^2 T_b}{4} \text{sinc}^2(fT_b) + \frac{A^2 T_b}{4} \text{sinc}^2(fT_b) \sum_{n=-\infty}^{\infty} e^{-j2\pi f n T_b}$$

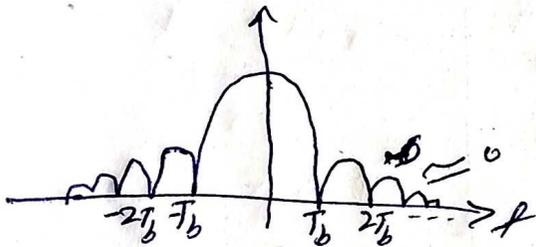
Now As per Poisson's formula:-

$$\sum_{n=-\infty}^{\infty} e^{-j2\pi f n T_b} = \frac{1}{T_b} \sum_{m=-\infty}^{\infty} \delta\left(f - \frac{m}{T_b}\right)$$

$$S(f) = \frac{A^2 T_b}{4} \text{sinc}^2(fT_b) + \frac{A^2}{4} \text{sinc}^2(fT_b) \sum_{m=-\infty}^{\infty} \delta\left(f - \frac{m}{T_b}\right)$$

$$S(f) = \frac{A^2 T_b}{4} \text{sinc}^2(f T_b) + \frac{A^2}{4} \delta(f)$$

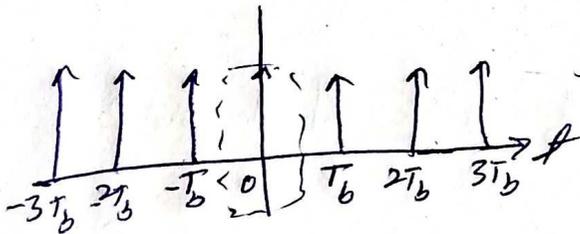
↑  
D.C. term which causes distortion in signal.



$\text{sinc}^2(f T_b) = 1$  at  $f = 0$  otherwise zero(0).

sinc function

} after multiplication of this function we get only output at zero.



impulse function